

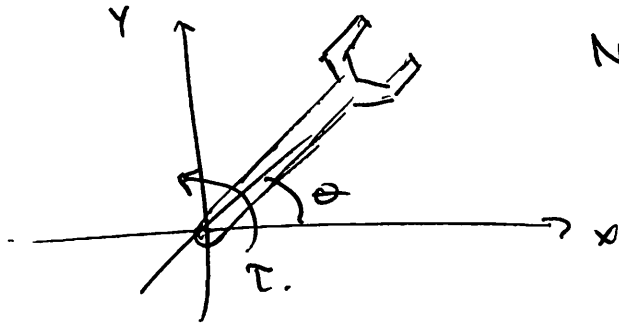
Recitation 11.

10-19-11

CT Feedback + Bode Plots.

Russ Tedrake

Consider controlling a simple robot arm.
(in the horizontal plane).



Note: Use robot arm prop in class.

Equations of motion are.

$$I \ddot{\theta} + b \dot{\theta} = \tau.$$

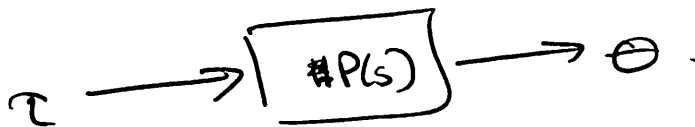
Inertia.
(e.g. $m l^2$)
always > 0 .

damping
(e.g. from friction)
always ≥ 0 .

motor torque

(motor current $\approx K$.
motor torque)

want to think of this as a system.



Q: Is it stable?

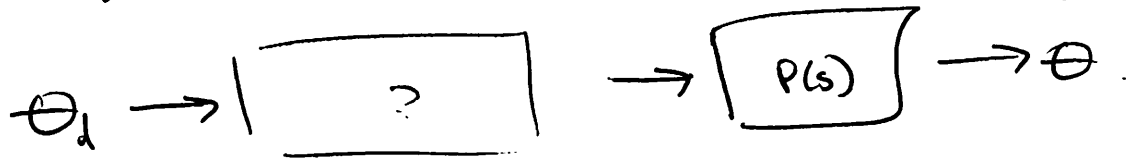
$$P(s) = \frac{1}{I s^2 + b s}$$

A: No: pole at zero.

physical demo:

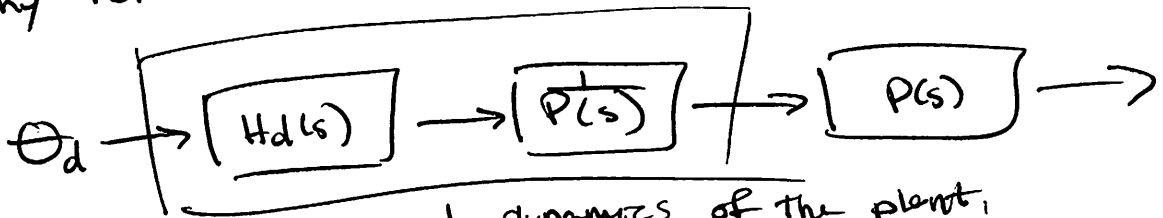
(I give a push, doesn't return to zero).

Goal: design a controller to control θ given θ_d .
(desired θ)

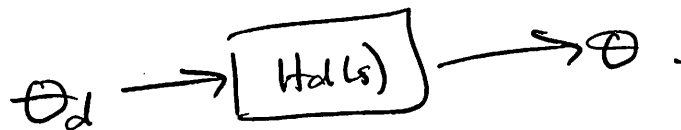


Philosophical aside:

Why not do:



e.g. cancel dynamics of the plant,
which should be
equivalent to.
then replace.

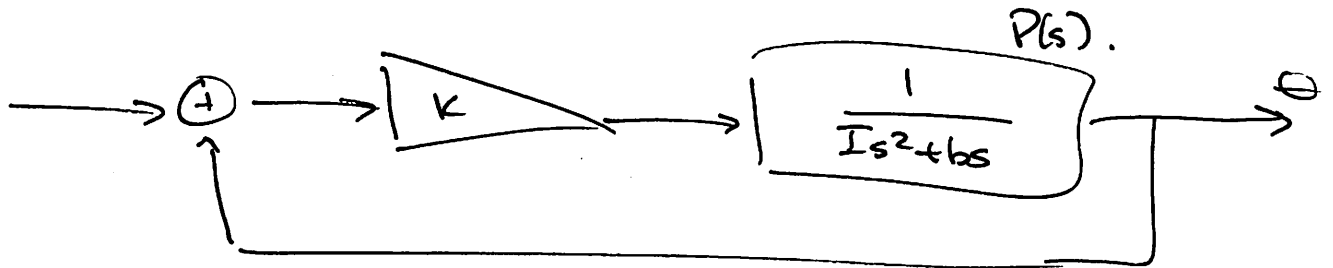


What if my model was a little wrong? ~~to~~

Some delay I didn't consider?

With feedback, relatively simpler controllers can
have relatively better performance (e.g. in terms
of robustness).

Idea. One: Proportional Control.

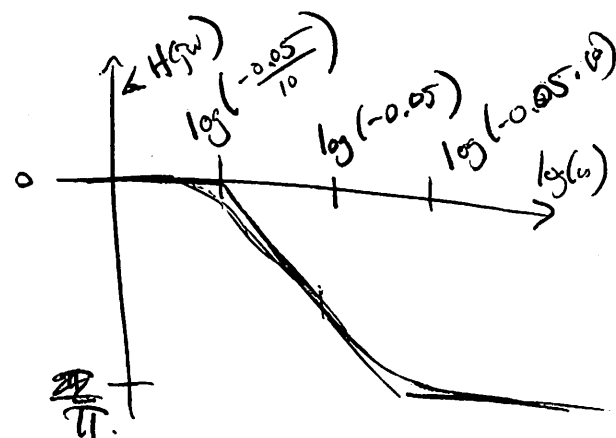
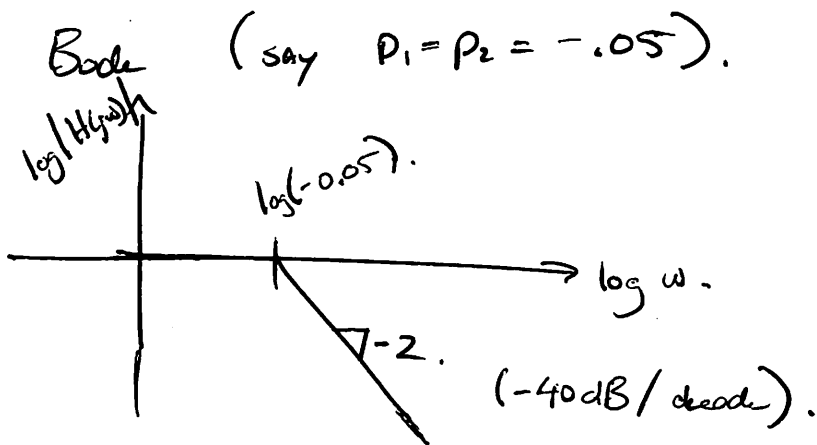
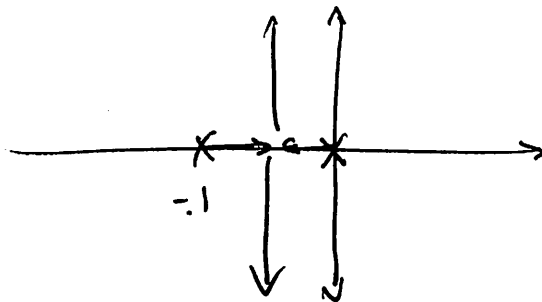


$$H(s) = \frac{\frac{K}{Is^2 + bs}}{1 + \frac{K}{Is^2 + bs}} = \frac{K}{Is^2 + bs + K}$$

Assume $m=1$, $l=1$, $b = \frac{1}{10}$, $g=10$.

$$\Rightarrow I = 1, b = \frac{1}{10}$$

Root locus.



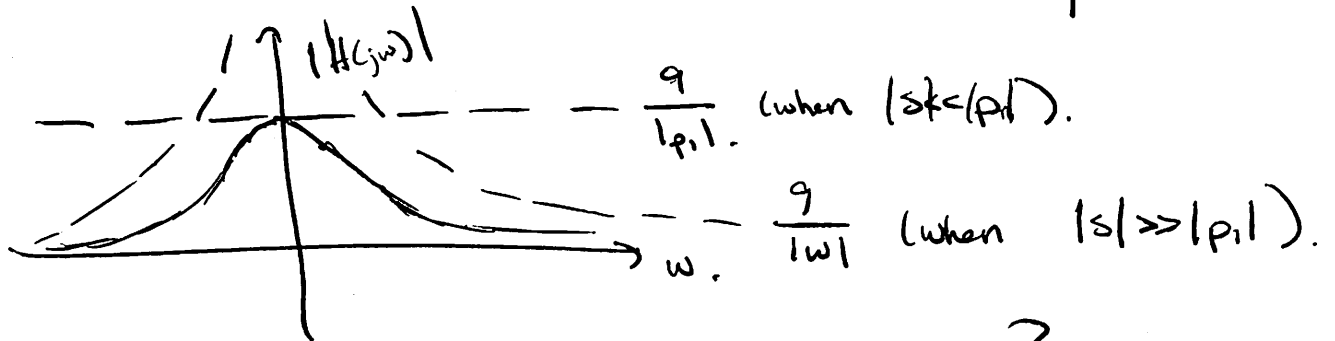
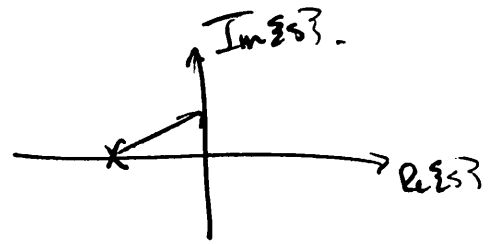
Main Problem: Slow response.

Q: Intuitively, how can we improve it? (Add a lead).

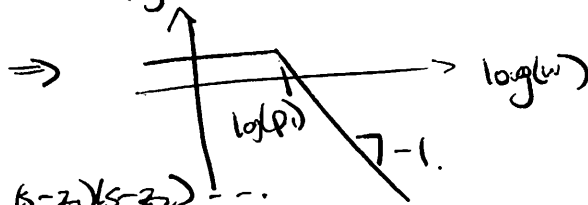
A: aka. derivative.

Bode approximations: (Recall from lecture).

Single pole: $H(s) = \frac{q}{s-p_1}$



$\log(|H(jw)|)$ where do they cross?
 $w = p_1$.



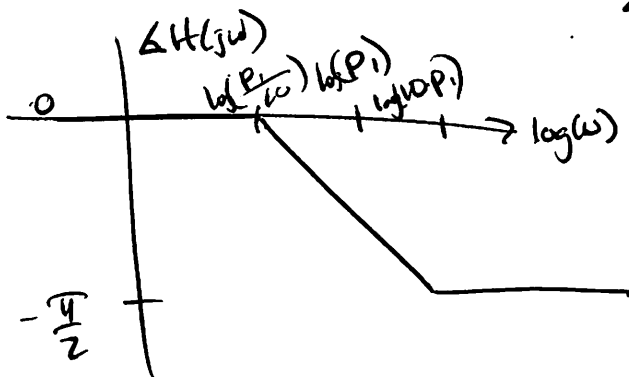
$$H(s) = K \frac{(s-z_1)(s-z_2)\dots}{(s-p_1)(s-p_2)\dots}$$

because $\log |H(jw)| = \log |K| + \sum \log |s-z_i| - \sum \log |s-p_i|$

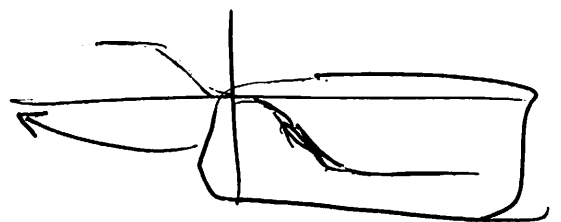
for phase. (don't need log, & already sums).

$$\Delta H(jw) = -\tan^{-1}\left(\frac{w}{p_1}\right)$$

$$\approx \begin{cases} 0 & w \ll |p_1| \text{ use } w \leq \frac{|p_1|}{10} \\ \text{linear} & \\ -\frac{\pi}{2} & w \gg |p_1| \text{ use } w \geq 10|p_1|. \end{cases}$$



Note: this is just for positive w.



Idea 2: Proportional Derivative Control. (PD).

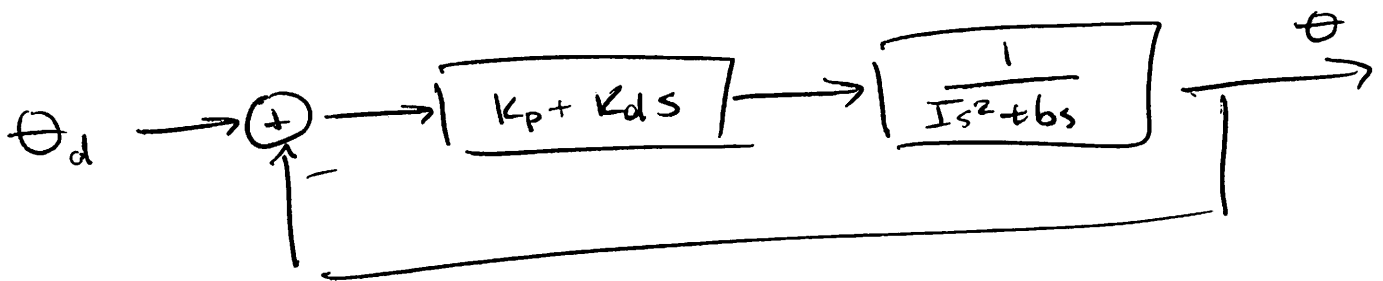
Proportional control was limited (rightmost pole could only be $-\frac{b}{2I}$)
+ large phase at high freq.

Big idea for faster response.
need to add a zero.

Derivative is my way to look into the future.

$$\tau = \underbrace{K_p (\theta_d - \theta)}_{\text{P control.}} + K_d (\dot{\theta}_d - \dot{\theta})$$

↑
"look into the future".

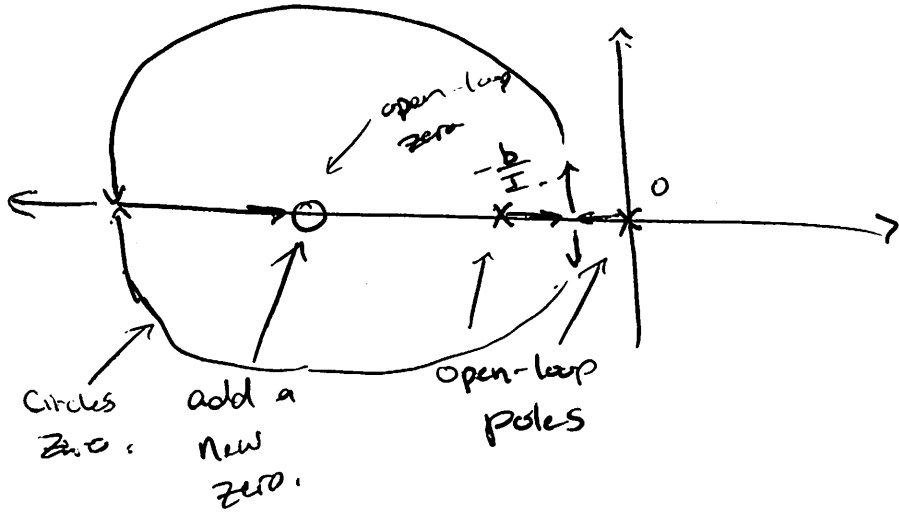


my claim: should get faster response.

$$H(s) = \frac{K_p + K_d s}{I s^2 + b s} = \frac{K_p + K_d s}{I s^2 + (b + K_d) s + K_p}$$
$$\frac{1 + \frac{K_p + K_d s}{I s^2 + b s}}$$

PD root locus:

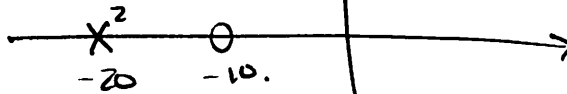
set $K_d = K_p/10$. (only vary one parameter).



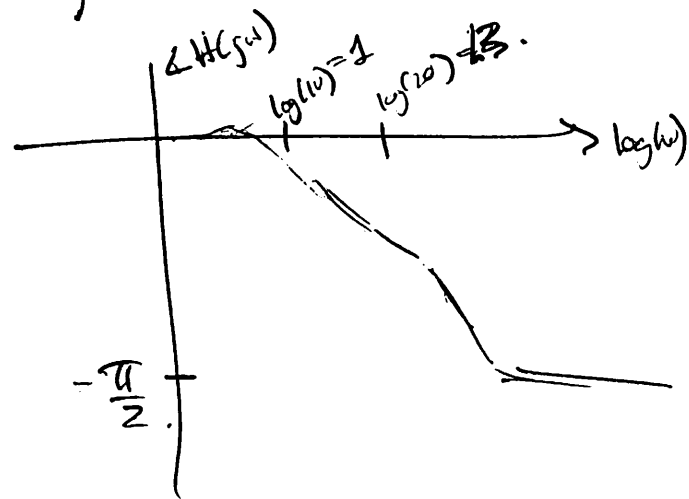
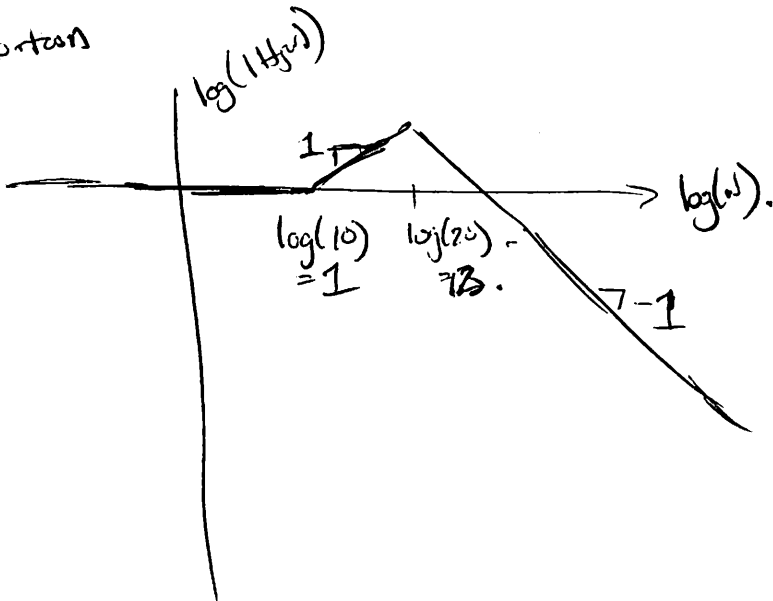
response can be much faster. (rightmost pole more negative).

Boice: (say for K_p such).
that I have ~~best~~ best response.

$K_p \approx 400$.



Cartan



Steady-state ~~error~~ error. (for unit-step input).

Final value theorem.

$$\lim_{t \rightarrow \infty} h(t) = \lim_{s \rightarrow 0} s H(s).$$

For ~~the~~ step response

$$\lim_{s \rightarrow 0} s H(s) \frac{1}{s} = \lim_{s \rightarrow 0} H(s).$$

often just $H(0)$.

Proportional control. steady-state:

$$H(s) = \frac{K}{Is^2 + bs + K}.$$

$$H(0) = \frac{K}{K} = 1 \quad \checkmark.$$

error = 0.

PD control steady-state:

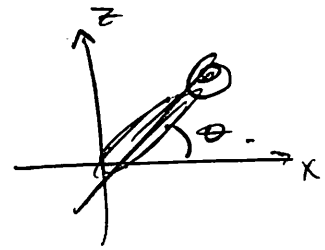
$$H(s) = \frac{K_p + K_d s}{Is^2 + (b + K_d)s + K_p}.$$

$$H(0) = \frac{K_p}{K_p} = 1 \quad \checkmark.$$

error = 0.

Robot in the vertical plane. (add gravity).

$$I\ddot{\theta} + b\dot{\theta} + mgl\cos\theta = \tau.$$

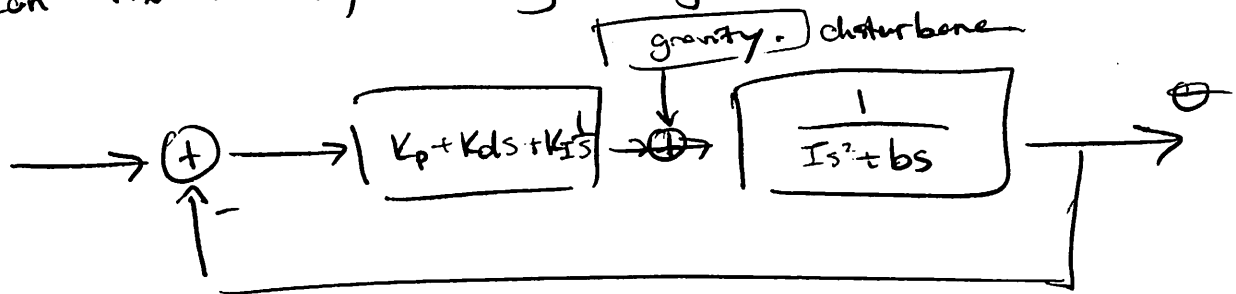


around $\theta=0$.

$$mgl\cos\theta \approx mgl.$$

will add steady-state error.

Can fix it by adding integral control.



PID control.

integral term will fix steady-state errors,
but makes things ~~more~~ slower. (pole at zero).

add derivative to speed back up.

Well-tuned P, I, and D gains (K_p, K_d, K_I)
can work very well for complex systems.